



## 4.2 Transmitter Power Control

### 4.2.1 General Parameters

Power control is applied for the TDD mode to limit the interference level within the system thus reducing the intercell interference level and to reduce the power consumption in the UE.

All codes within one timeslot allocated to the same CCTrCH use the same transmission power, in case they have the same spreading factor.

**Table 1: Transmit Power Control characteristics**

	Uplink	Downlink
<b>Power control rate</b>	Variable 1-7 slots delay (2 slot SCH) 1-14 slots delay (1 slot SCH)	Variable, with rate depending on the slot allocation.
<b>Step size</b>		1, 2, 3 dB
<b>Remarks</b>	All figures are without processing and measurement times	Within one timeslot the powers of all active codes may be balanced to within a range of 20 dB

### 4.2.2 Uplink Control

#### 4.2.2.1 General Limits

By means of higher layer signalling, the Maximum\_Allowed\_UL\_TX\_ power for uplink may be set to a value lower than what the terminal power class is capable of. The total transmit power shall not exceed the allowed maximum. If this would be the case, then the transmit power of all uplink physical channels in a timeslot is reduced by the same amount in dB.

#### 4.2.2.2 PRACH

The transmit power for the PRACH is set by higher layers based on open loop power control as described in TS 25.331.

#### 4.2.2.3 DPCH, PUSCH

##### 4.2.2.3.1 Gain Factors

Two or more transport channels may be multiplexed onto a CCTrCH as described in [9]. These transport channels undergo rate matching which involves repetition or puncturing. This rate matching affects the transmit power required to obtain a particular  $E_b/N_0$ . Thus, the transmission power of the CCTrCH shall be weighted by a gain factor  $\beta$ .

There are two ways of controlling the gain factors for different TFC's within a CCTrCH transmitted in a radio frame:

- $\beta$  is signalled for the TFC, or
- $\beta$  is computed for the TFC, based upon the signalled settings for a reference TFC.

Combinations of the two above methods may be used to associate  $\beta$  values to all TFC's in the TFCS for a CCTrCH. The two methods are described in sections 4.2.2.3.1.1 and 4.2.2.3.1.2 respectively. Several reference TFC's for several different CCTrCH's may be signalled from higher layers.

The weight and gain factors may vary on a radio frame basis depending upon the current SF and TFC used. The setting of weight and gain factors is independent of any other form of power control. That means that the transmit power  $P_{UL}$  is calculated according to the formula given below in section 4.2.2.3.2 and then the weight and gain factors are applied on top of that, cf. [10].

#### 4.2.2.3.1.1 Signalled Gain Factors

When the gain factor  $b_j$  is signalled by higher layers for a certain TFC, the signalled values are used directly for weighting DPCH(s). Exact values are given in [10].

#### 4.2.2.3.1.2 Computed Gain Factors

The gain factor  $b_j$  may also be computed for certain TFCs, based on the signalled settings for a reference TFC:

Let  $b_{ref}$  denote the signalled gain factor for the reference TFC. Further, let  $b_j$  denote the gain factor used for the  $j$ -th TFC.

Define the variable: 
$$K_{ref} = \sum_i RM_i \cdot N_i$$

where  $RM_i$  is the semi-static rate matching attribute for transport channel  $i$ ,  $N_i$  is the number of bits output from the radio frame segmentation block for transport channel  $i$  and the sum is taken over all the transport channels  $i$  in the reference TFC.

Similarly, define the variable 
$$K_j = \sum_i RM_i \cdot N_i$$

where the sum is taken over all the transport channels  $i$  in the  $j$ -th TFC.

Moreover, define the variable 
$$L_{ref} = \sum_i \frac{1}{SF_i}$$

where  $SF_i$  is the spreading factor of DPCH  $i$  and the sum is taken over all DPCH  $i$  used in the reference TFC.

Similarly, define the variable 
$$L_j = \sum_i \frac{1}{SF_i}$$

where the sum is taken over all DPCH  $i$  used in the  $j$ -th TFC.

Then the variable  $A_j$ , called the nominal power relation for TFC  $j$ , is computed as:

$$A_j = \sqrt{\frac{L_j}{L_{ref}}} \times \sqrt{\frac{K_{ref}}{K_j}}$$

The gain factors  $b_j$  for the  $j$ -th TFC are then computed as follows:

- If  $A_j > 1$ , then  $b_j$  is the largest quantized  $b$ -value, for which the condition  $b_j \leq 1 / A_j$  holds.
- If  $A_j \leq 1$ , then  $b_j$  is the smallest quantized  $b$ -value, for which the condition  $b_j \geq 1 / A_j$  holds.

The quantized  $\beta$ -values are given in [10].

#### 4.2.2.3.2 Power Control Loop

After the synchronisation between UTRAN and UE is established, the UE transits into open-loop transmitter power control (TPC).

The transmitter power setting for each uplink DPCH in one CCTrCH of UE shall be calculated by the following equation:

$$P_{UE} = \alpha P_{CPCH} + (1 - \alpha) L_0 + I_{BTS} + SIR_{TARGET} + \text{Constant value}$$

where

$P_{UE}$ :	Transmitter pPower setting level in dBm, cf. section "Combination of physical channels in uplink" in [10]; This value corresponds to a particular CCTrCH (due to CCTrCH-specific $SIR_{TARGET}$ ) and a particular timeslot (due to possibly timeslot-specific $\alpha$ and $I_{BTS}$ ).
$L_{P-CCPCH}$ :	Measure representing path loss in dB (reference transmit power is broadcast on BCH).
$L_0$ :	Long term average of path loss in dB.
$I_{BTS}$ :	Interference signal power level at cell's receiver in dBm, which is broadcast on BCH.
$\alpha$ :	$\alpha$ is a weighting parameter which represents the quality of path loss measurements. $\alpha$ may be a function of the time delay between the uplink time slot and the most recent down link time slot containing a physical channel that provides the beacon function, see [8]. $\alpha$ is calculated at the UE. An example for calculating $\alpha$ as a function of the time delay is given in annex A.1.
$SIR_{TARGET}$ :	Target SNR in dB. A higher layer outer loop adjusts the target SIR.
Constant value:	This value shall be set by higher Layer (operator matter). and is broadcast on BCH.

If the midamble is used in the evaluation of  $L_{P-CCPCH}$  and  $L_0$ , and the Tx diversity scheme used for the P-CCPCH involves the transmission of different midambles from the diversity antennas, the received power of the different midambles from the different antennas shall be combined prior to evaluation of these variables.

#### 4.2.2.3.34 Out of synchronisation handling

UE shall shut off the uplink transmission if the following criteria is fulfilled:

- the UE estimates the received dedicated channel burst quality over the last [160] ms period to be worse than a threshold  $Q_{out}$ . This criterion is never fulfilled during the first [160] ms of the dedicated channel's existence.  $Q_{out}$  is defined implicitly by the relevant tests in TS 25.102;
- if the UE detect the beacon channel reception level [10 dBm] above the handover triggering level, then the UE uses [320] ms estimation period for the burst quality evaluation.

UE shall resume the uplink transmission if the following criteria is fulfilled:

- the UE estimates the burst reception quality over the last [160] ms period to be better than a threshold  $Q_{in}$ . This criterion is always fulfilled during the first [160] ms of the dedicated channel's existence.  $Q_{in}$  is defined implicitly by the relevant tests in TS 25.102.