Meeting \#15<br>22-26 ${ }^{\text {th }}$, August, 2000<br>Berlin, Germany

| Agenda Item : | 7 |
| :--- | :--- |
| Source : | Motorola |
| Title : | Verification algorithm for closed loop transmit diversity mode 2 |
| Document for : | Discussion |

## Introduction

The RAN1 specifications currently contain an example verification algorithm for transmit diversity mode 1 . Although, the example algorithm only applies to mode 1, the general principle of verification can be extended for transmit diversity mode 2.

This document describes a general verification algorithm for transmission diversity mode 2 . Simulation results are given in order to compare the performance of mode 2 with and without verification.

## 1 Generic Extension of verification for mode 2

For the current mode 2 specification, the same pilot sequence is transmitted on both antennas for DPCCH. Therefore, we obtain the equivalent channel estimation which corresponds to the channel seen by the UE receiving antenna (equivalent channel seen by the DPCCH) :

$$
\vec{h}^{(d)}=\gamma\left(\beta_{1} \vec{h}_{1}+\beta_{2} \vec{h}_{2}\right)+\vec{b}^{(d)} \text { Eqn. (1) }
$$

where $\gamma^{2}$ is the DPCCH Pilot SNIR/CPICH SNIR, Signal to Noise plus Interference Ratio and $\beta_{1}, \beta_{2}$ are the applied coefficients on the antennas at the UTRAN. Furthermore, in all cases, we have channel estimations made on CPICH Pilots for each antenna:

$$
\begin{array}{ll}
\vec{h}_{1}^{(p)}=\vec{h}_{1}+\vec{b}_{1}^{p} & \text { Eqn (2) } \\
\vec{h}_{2}^{(p)}=\vec{h}_{2}+\vec{b}_{2}^{p} & \text { Eqn (3) }
\end{array}
$$

where $\vec{h}_{i}^{(p)}$ is the estimated channel vector using the CPICH, $\vec{h}_{i}$ is the actual channel vector, and $\vec{b}_{i}{ }^{p}$ is the noise vector, from the $i$-th antenna.

At the receiver, verification consists in choosing a pair of applied coefficients at the UTRAN, $\left(\hat{\beta_{1}}, \hat{\beta_{2}}\right)$, which fits best to the actual applied one (c.f. signaling errors in the annex A in [2]).

One possible way of implementing verification for mode 2 is by choosing $\left(\hat{\beta}_{1}, \hat{\beta}_{2}\right)$ from the whole set of possibilities $T=\left\{\alpha_{1}, \alpha_{2}\right\}$, using the logarithmic form, of the following decision rule:

$$
\begin{equation*}
\left(\hat{\beta_{1}}, \hat{\beta_{2}}\right)=\arg \left\langle\max _{\alpha_{1}, \alpha_{2} \in T}\left\{\ln \left(\hat{p}\left(\alpha_{1}, \alpha_{1}\right)\right)+\ln \left(\bar{p}\left(\alpha_{1}, \alpha_{2}\right)\right)\right\}\right\rangle \quad \text { Eqn. } \tag{4}
\end{equation*}
$$

where , in normal operation the a priori probability $\bar{p}\left(\alpha_{1}, \alpha_{2}\right)$ for the selected antenna coefficients pair (asked word at the UE) is assumed to be $96 \%$ (assuming there are $4 \%$ of errors in the feedback channel).

$$
\ln \left(\hat{p}\left(\hat{\beta_{1}}, \hat{\beta_{2}}\right)\right)=-\left(\sum_{i=1}^{\text {Npath } \left.\frac{\mid h_{i}^{(d)}-\gamma\left(\hat{\beta_{1}} h_{1, i}^{(p)}+\hat{\beta}_{2} h_{2, i}^{(p)}\right)^{2}}{\sigma_{i}^{2}\left(1+\gamma^{2}\left(\left|\hat{\beta}_{1}\right|^{2}+\left|\hat{\beta}_{2}\right|^{2}\right)\right)}\right) \text { Eqn. (5) }}\right.
$$

## 2 Simulations assumptions

| Parameter | Explanation/Assumption |  |
| :---: | :---: | :---: |
| Service | Speech 12.2 kbps |  |
| Chip Rate | 3.84 Mcps |  |
| Closed loop Power Control | OFF |  |
| Channel Estimation | PILOT ; DPCCH if verification OFF , CPICH if verification ON |  |
| FBI error rate | 0\%, 4\%,and 10\% |  |
| Number of samples per chip | 1 |  |
| FBI rate | 1500 bps |  |
| Propagation Conditions | ETRP A, ETRV A |  |
| Number of Rake Fingers | 2 for ETRP and 4 for ETRV |  |
| Downlink Physical Channels and Power Levels | CPICHP_Ec1/lor | $=-13 \mathrm{~dB}$ |
|  | CPICHP_Ec2/lor | $=-13 \mathrm{~dB}$ |
|  | PCCPCH_Ec1/lor | $=-15 \mathrm{~dB}$ |
|  | PCCPCH_Ec2/lor | $=-15 \mathrm{~dB}$ |
|  | PICH Ec/lor | $=-15 \mathrm{~dB}$ |
|  | SCH_Ec/lor | $\begin{aligned} & =-12 \mathrm{~dB} \text { (Combined energy of Primary and } \\ & \text { Secondary SCH) } \end{aligned}$ |
| $\hat{I}_{o r} / I_{o c}$ value | 6 dB |  |

Table 1. Simulation parameters

## 3 Simulation results for mode 2 with verification

The following figures show the performance of the generic verification algorithm for transmission diversity mode 2. The performance is shown for the following channel models:

- ETRP, $10 \mathrm{~km} / \mathrm{h}$
- ETRP, $20 \mathrm{~km} / \mathrm{h}$
- ETRV, 50km.h
and for the following feedback error rates:
- $0 \%$
- $4 \%$
- $10 \%$

At a BLER of $10^{-2}$ and for a nominal feedback error rate of $4 \%$, verification for mode 2 gives a gain of 0.7 dB for all of the channel models considered.

### 3.1 Performance for ETRP A 10 km/h





### 3.2 Performance for ETRP A 20 km/h





### 3.3 Performance for ETRV A 50 km/h





## 4 Conclusions

- A generic algorithm for verification for transmission diversity mode 2 has been defined and has been validated through simulations.: At a BLER of $10^{-2}$ and for a nominal feedback error rate of $4 \%$, verification for mode 2 gives a gain of 0.7 dB for all mobile speeds up to $50 \mathrm{~km} / \mathrm{h}$.
- The verification algorithm described does not require any changes to the current RAN1 specification. Therefore, the algorithm could be implemented within a release 99 UE.
- For release 2000, the verification algorithm for mode2 could be described in 25.214 annex A(informative) that already contains the verification algorithm for mode1.
- The performance benefits of applying verification for transmit diversity mode 1 and 2 should be considered by RAN4 in setting minimum performance requirements for release 2000.


## 5 References

[1] R1-99c86, Modified closed loop modes for WCDMA, TI, Aug $24^{\text {th }}$, TSG-RAN WG1 meeting \#7, Aug $30^{\text {th }}$ Sep $3{ }^{\text {rd }}$, Hanover, Germany
[2] 3G TS 25.214 v3.2.0, TSG RAN, Physical layer procedures (FDD), 2000-03.

